

Audio Engineering Society

Convention Paper

Presented at the 120th Convention 2006 May 20–23 Paris, France

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Improved Noise Weighting in CELP Coding of Speech - Applying the Vorbis Psychoacoustic Model To Speex

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ABSTRACT

One key aspect of the CELP algorithm is that it shapes the coding noise using a simple, yet effective, weighting filter. In this paper, we improve the noise shaping of CELP using a more modern psychoacoustic model. This has the significant advantage of improving the quality of an existing codec without the need to change the bit-stream. More specifically, we improve the Speex CELP codec by using the psychoacoustic model used in the Vorbis audio codec. The results show a significant increase in quality, especially at high bit-rates, where the improvement is equivalent to a 20% reduction in bit-rate. The technique itself is not specific to Speex and could be applied to other CELP codecs.

1. INTRODUCTION

When the Code-Excited Linear Prediction (CELP) [1] scheme was originally proposed, one key aspect of the algorithm was how the noise was shaped using a simple, yet effective, weighting filter. Since then, audio coding advances have provided significantly better psychoacoustic models for shaping coding noise. However, psychoacoustic modeling in CELP has remained essentially the same. In this paper, we propose to improve the quality of an existing CELP codec by using better noise

shaping, without modifying the bit-stream of the codec. More specifically, we improve the Speex¹ CELP codec by using a psychoacoustic model derived from the Vorbis² audio codec.

Speex is an open-source multi-rate CELP codec supporting both narrowband and wideband speech. Unlike most current CELP codecs, it uses a 3-tap pitch predictor and sub-vector innovation quantization. Vorbis

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²http://www.vorbis.org/

is a high-quality open-source audio codec designed for music and uses the modified discrete cosine transform (MDCT). Both Speex and Vorbis are developed within the Xiph.Org Foundation.

This paper is organized as follows. Section 2 introduces the Speex codec used in this work. Section 3 then describes the psychoacoustic model used by the Vorbis codec. The application of that model to Speex is described in Section 4. Results are presented in Section 5 with a discussion in Section 6.

2. THE SPEEX CODEC

Speex is an open-source codec based on the Code-Excited Linear Prediction (CELP) algorithm. It is targeted mainly towards voice over IP (VoIP) applications so it is designed to be robust to lost packets. Speex supports multiple bit-rate, ranging from 2.15 kbps to 24.6 kbps in narrowband (8 kHz) operation and from 3.95 kbps to 42.2 kbps in wideband (16 kHz) operation. Some additional features in Speex are:

- Embedded wideband coding
- Variable bit-rate (source controlled)
- Voice activity detection (VAD) and discontinuous transmission (DTX)
- Variable search complexity

The Speex bit-stream was frozen in March 2003 with the release of version 1.0. However, since there is no "bit-exact" specification, it is still possible to improve the quality of the encoder as long as no modification is required on the decoder side.

2.1. Perceptual Weighting

In order to maximize speech quality, CELP codecs minimize the mean square of the error (noise) in the perceptually weighted domain. This means that a perceptual noise weighting filter W(z) is applied to the error signal in the encoder. In most CELP codecs, W(z) is a pole-zero weighting filter derived from the linear prediction coefficients (LPC), generally using bandwidth expansion. Let the spectral envelope be represented by the synthesis filter 1/A(z), CELP codecs typically derive the noise weighting filter as:

$$W(z) = \frac{A(z/\gamma_1)}{A(z/\gamma_2)} \tag{1}$$

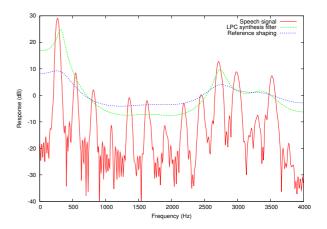


Figure 1: Standard noise shaping in CELP. Arbitrary y-axis offset.

where $\gamma_1 = 0.9$ and $\gamma_2 = 0.6$ in the Speex reference implementation.

The weighting filter is applied to the error signal used to optimize the codebook search through analysis-by-synthesis (AbS). This results in a spectral shape of the noise that tends towards 1/W(z). While the simplicity of the model has been an important reason for the success of CELP, it remains that W(z) is a very rough approximation for the perceptually optimal noise weighting function. Fig. 1 illustrates the noise shaping that results from Eq. 1. Throughout this paper, we refer to W(z) as the noise weighting filter and to 1/W(z) as the noise shaping filter (or curve).

2.2. Narrowband Encoder Structure

In narrowband, Speex frames are 20 ms long (160 samples) and are subdivided in 4 sub-frames of 5 ms each (40 samples). For most narrowband bit-rates (8 kbps and above), the only parameters encoded at the frame level are the Line Spectral Pairs (LSP) and a global excitation gain g_{frame} , as shown in Fig. 2. All other parameters are encoded at the sub-frame level.

Linear prediction analysis is performed once per frame using an asymmetric Hamming window centered on the fourth sub-frame. The linear prediction coefficients (LPC) are converted to line spectral pairs (LSP) and vector-quantized using 30 or 18 bits (depending on the bit-rate used). To make Speex more robust to packet loss, no prediction is applied on the LSP coefficients

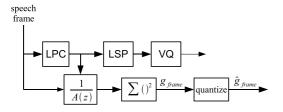


Figure 2: Frame open-loop analysis

prior to quantization. For each sub-frame, the LSP coefficients are interpolated linearly based on the current and past quantized LSP coefficients and converted back to the LPC filter $\hat{A}(z)$. The non-quantized interpolated filter is denoted A(z).

The analysis-by-synthesis (AbS) encoder loop is described in Fig. 3. There are three main aspects where Speex significantly differs from most other CELP codecs. First, while most recent CELP codecs make use of fractional pitch estimation [2] with a single gain, Speex uses an integer to encode the pitch period, but uses a 3-tap predictor [3] (3 gains). The adaptive codebook contribution $e_a[n]$ can thus be expressed as:

$$e_a[n] = g_0e[n-T-1] + g_1e[n-T] + g_2e[n-T+1]$$
 (2)

where g_0 , g_1 and g_2 are the jointly quantized pitch gains and e[n] is the codec excitation memory.

Many current CELP codecs use moving average (MA) prediction to encode the fixed codebook gain. This provides slightly better coding at the expense of introducing a dependency on previously encoded frames. A second difference is that Speex encodes the fixed codebook gain as the product of the global excitation gain g_{frame} with a sub-frame gain corrections g_{subf} . This increases robustness to packet loss by eliminating the inter-frame dependency. The sub-frame gain correction is encoded before the fixed codebook is searched (not closed-loop optimized) and uses between 0 and 3 bits per sub-frame, depending on the bit-rate.

The third difference is that Speex uses sub-vector quantization of the innovation (fixed codebook) signal instead of an algebraic codebook. Each sub-frame is divided into sub-vectors of lengths ranging between 5 and 20 samples. Each sub-vector is chosen from a bitrate-dependent codebook and all sub-vectors are concatenated to form a sub-frame. As an example, the 3.95 kbps mode uses

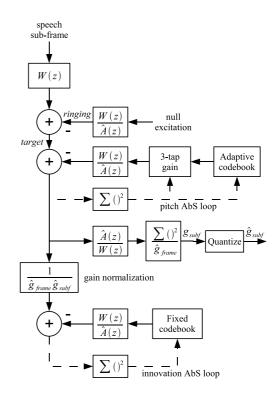


Figure 3: Analysis-by-synthesis closed-loop optimization on a sub-frame.

a sub-vector size of 20 samples with 32 entries in the codebook (5 bits). This means that the innovation is encoded with 10 bits per sub-frame, or 2000 bps. On the other hand, the 18.2 kbps mode uses a sub-vector size of 5 samples with 256 entries in the codebook (8 bits), so the innovation uses 64 bits per sub-frame, or 12800 bps.

3. THE VORBIS PSYCHOACOUSTIC MODEL

The masking model we use in this paper is based on elements of the psychoacoustic model of the Vorbis opensource audio codec, specifically noise masking. Noise masking in the Vorbis codec is implemented in a conceptually similar manner to the Spectral Flatness Measure introduced by Johnston [4]. A geometric median and envelope follower are constructed by smoothing the log spectrum with a sliding window of approximately one Bark. The distance between the two curves provides a tonality estimate for a given band of the spectrum. The envelope curve is companded according to the distance in order to directly compute a *noise mask* curve that is used for spectral weighting.

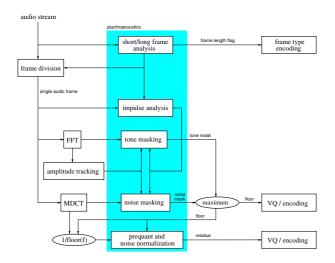


Figure 4: Block diagram of the Vorbis encoder applied to a mono input stream. The shaded block contains the psychoacoustic elements.

3.1. Psychoacoustics Basic Ideas

Audio coding uses psychoacoustics in two ways. The first is a separable mechanism that quantifies the relative importance of specific audio features so that the encoding may be weighted toward representing the most important. This analytical optimization is entirely encoderside and although it operates within the framework of the codec design, it tends not to be part of a codec specification. The metrics and decision logic may change without affecting the compatibility of the bit-stream. In this role, psychoacoustics are essential to efficiency but optional to correct mechanical operation. Most codecs, including Vorbis and Speex fall in this category.

Secondly, every audio codec also relies on some amount of psychoacoustic design hardcoded into the codec specification. Such hardwired psychoacoustics are part of the required mechanisms of both the encoder and decoder. A number of contemporary codecs dispense with the former *optional* psychoacoustics altogether, relying entirely on the hardwired psychoacoustics of the codec's explicit design. In such codecs, like the transform-coded excitation (TCX) [5] algorithm, it is not possible to change the psychoacoustic model without changing the bit-stream.

3.2. The Vorbis Model

The Vorbis codec relies on encode-side psychoacoustic heuristics in order to produce effective bit-streams, as illustrated in Fig. 4. From each frame's audio spectrum, it computes a *floor* curve which is used as a unit-resolution whitening filter, that is, the floor is used to normalize the spectrum such that a direct linear quantization of the normalized spectrum is perceptually appropriate. Thus, the floor performs a similar function to the conceptually simpler perceptual weighting filter used in Speex and other CELP codecs. However, in the case of Vorbis, the floor curve is required for decoding, whereas CELP *merely* uses the weighting filter for optimizing the search for the best entries in the codebooks.

The frame-by-frame psychoacoustic metrics used by the Vorbis encoder to compute the floor fall into four main categories: tone masking, noise masking, noise normalization, and impulse analysis.

3.2.1. Tone Masking

Tone masking in the reference Vorbis encoder is a simple tone-tone masking engine which produces a single spectral curve per frame. This curve represents the threshold of perception, per spectral line, for a pure tone in the current spectral frame. The curve is computed by superposition of tonal masking curves such as those originally presented by Ehmer [6]. Note that the reference encoder uses curves measured independently by Xiph.Org.

Tone-tone masking analysis and the tonal masking curve are of only marginal direct importance in the modern Vorbis codec; most commercially produced audio programmes have seen deep single- and multi-band compression such that there is insufficient dynamic depth for tone masking to result in much savings.

3.2.2. Noise Masking

Noise masking in Vorbis is something of a historical misnomer; although it began as a simple noise-noise masking mechanism analogous to the tone-tone masking above, in time it has become more concerned with constructing a curve that represents the approximate envelope of noise energy in the spectrum. This curve is computed from geometric median and envelope followers constructed by smoothing the log spectrum with a sliding window of approximately one Bark. The envelope curve is companded according to the distance between the envelope and the mean; greater distances imply greater tonality and lower distances imply greater noisiness. Accordingly, the amplitude of this mask curve is depressed in areas of greater tonality. Vorbis then adds a hardwired bias curve (the noise offset) to the companded envelope, producing the final *noise mask* curve.

3.2.3. Noise Normalization

Noise normalization is conceptually part of analyzing and handling noise energy as described under noise masking, but it is handled in a separate step. The purpose of normalization is to preserve approximate wideband energy through quantization, especially at very low S/N ratios, where large portions of the spectrum may otherwise collapse toward silence. It may be viewed as a mechanism that preserves the intent of noise masking through the quantization process. The need for the specific implementation of noise normalization in the Vorbis encoder is a consequence of the nature of a frequency domain codec.

3.2.4. Impulse Analysis

Impulse analysis refers to several metrics that characterize highly temporally localized events in an audio frame, such as a sudden impulse or attack in the audio. Earlier stages in the Vorbis encoder analyze attacks and preecho potential to determine when to switch between short and long frames, reusing these metrics in the frame-internal coding algorithm. Frame analysis also inspects audio frames for an impulse-train-like nature, that is, audio resembling a filter being driven by an impulse train; such audio tends to have a particularly regular harmonic structure into the high harmonics. Voice is the most obvious example of this variety of audio.

As with noise normalization, this additional impulse analysis is necessitated by the nature of a transform codec. Sudden impulses and the characteristic tight *rasp* of impulsive audio are features not compactly representable in the frequency domain. Naive quantization causes narrow events to smear in time and this loss of temporal resolution is perceptually obvious. Impulse analysis is used to improve representation of non-sinusoidal, non-random-noise content.

3.2.5. Floor Construction

The final floor curve in vorbis is created from the maximum of a direct superposition of the tone mask and noise mask curves. The floor curve is removed from the MDCT spectrum of a given audio frame, thus whitening the spectrum and resulting in *spectral residual* values. The floor and residue are then coded via vector quantization.

4. APPLICATION TO CELP CODING

Speex does not use multiple blocksizes and for that reason, the block-switching and analysis of the Vorbis codec

is irrelevant. In addition, Speex natively represents timelocalized events and impulsive audio characteristics very well. The impulse analysis as implemented in the Vorbis codec is specific to transform codecs and as such is also not relevant to Speex.

Similarly, noise normalization also addresses a need highly specific to the frequency transform domain encoder. Noise normalization as realized and used by the Vorbis encoder prevents gross wideband energy inflation or collapse due to naive quantization, a situation from which Speex is relatively well proofed by virtue of being an LPC-based codec.

Tone and noise masking as described in Section 3.2 retain relevance in the context of the Speex codec. Experience with the Vorbis encoder indicates that noise masking is responsible for the greatest bulk of useful bit-rate savings. For purposes of initial experimentation, we thus concentrate on implementing noise masking alone in the Speex codec.

One of the assumptions made in the Vorbis codec is that the quantization noise is entirely masked. After all, it must be for the codec to achieve transparency or neartransparency. This assumption leads to using a noise weighting curve that is very close to the masking curve, which means a constant (negative) noise-to-mask ratio. However, the assumption is not valid for Speex because there are simply not enough bits available for the noiseto-mask ratio to be negative (or zero) at all frequencies. This means that the quantization noise is always audible to some extent. Using the masking curve directly for noise weighting in Speex would results in over-emphasis of the noise in the high-energy regions of the spectrum (typically low frequencies). For that reason, it is not desirable to have a constant (positive) noise-to-mask ratio and the masking curve used by Vorbis needs to be modified.

In order to obtain good results with the Speex codec, we need to compress the dynamic range of the masking curve. It was determined empirically that the optimal companding consists of applying an exponent of 0.6 to the masking curve computed by Vorbis. This value was found to be suitable (near-optimal) for all bit-rates.

4.1. Masking Curve to Weighting Filter

In the integration of the new psychoacoustic model in Speex, it is desirable to keep the same pole-zero formu-

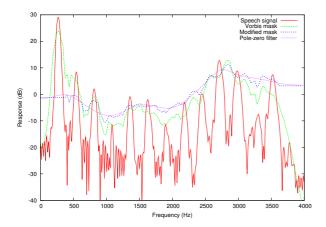


Figure 5: Modifications to the Vorbis mask. Arbitrary y-axis offset.

lation for the noise shaping filter:

$$\frac{1}{W(z)} = \frac{W_n(z)}{W_d(z)} \tag{3}$$

For that reason, the frequency-domain noise shaping curve is converted to a 10^{th} order pole-zero filter using a two-step method. The filter denominator $W_d(z)$ is first obtained by transforming the masking curve into auto-correlation coefficients using an inverse FFT and then applying the Levinson-Durbin algorithm.

The filter numerator $W_n(z)$ is then estimated based on the error between the all-pole model and the real masking curve. In practice, $W_n(z)$ is nearly flat because $W_d(z)$ alone is able to provide a very good approximation of the real curve.

Transforming the masking curve to a pole-zero model not only makes the implementation easy, but it preserves the efficiency of the CELP analysis-by-synthesis (AbS) codebook search. In order to limit the complexity, the masking curve is computed only once every frame. For each sub-frame, the curve is linearly interpolated and converted to a pole-zero filter.

4.2. Complexity Reduction

The computation and conversion of the masking curve described above tends to increases the complexity of the Speex codec. To compensate for that, we propose three methods to minimize the impact of the proposed psychoacoustic model:

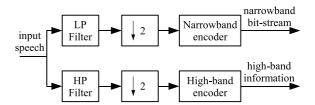


Figure 6: Speex wideband encoder

1. Not using the filter numerator

Because the weighting filter numerator $W_n(z)$ generally has little effect, it can be omitted without having a significant impact on quality. This decreases the complexity of converting the masking curve to a pole-zero filter, while simplifying the weighting filter.

2. Setting the denominator to be the same as the synthesis filter

By forcing $W_d(z) = A(z)$, it is also possible to reduce the complexity. While the complexity reduction in the conversion is smaller than in method 1, the main advantage lies in the fact that the most commonly used filter in the encoder (Fig. 3) simplifies to $\frac{W(z)}{A(z)} = \frac{1}{W_n(z)}$. This reduces the complexity of the encoder by approximately one million operations (add or multiply) per second.

3. Using a frame-constant numerator

By using method 2 and keeping the same numerator $W_n(z)$ for whole frames, we can reduce the cost of converting the filters by a factor of four. The resulting complexity becomes similar to that of the reference encoder, while still providing a quality improvement.

4.3. Application to Wideband Coding

To facilitate inter-operability with the public switched telephone network (PSTN), Speex encodes wideband speech using the Sub-Band CELP (SB-CELP) technique. This consists of splitting the signal in two bands using a quadrature mirror filter (QMF), as shown in Fig. 6. The lower band (0-4 kHz) is encoded using the narrow-band encoder, while the higher band is encoded using a pitch-less version of CELP (denoted HF-CELP). No pitch prediction is used for the higher band because the

spectral folding caused by the QMF makes the signal non-periodic.

Only three parameters/vectors are transmitted: the highband LSP parameters, sub-frame gain corrections and the innovation signal. We use 8 LSPs, jointly quantized with 12 bits. The gain corrections are computed based on the ratio between high-band to low-band excitation and correcting for the LPC response difference at the cutoff frequency (4 kHz). The innovation can be encoded using 8, 4, or 0 kbps. In the lowest bit-rate, only the shape of the spectrum is preserved and the excitation is a frequency-aliased version of the narrowband part. This is done using a technique conceptually similar to [7] and requires only 1.8 kbps to transmit the higher band.

Because of the embedded structure, no additional work is necessary to make the proposed psychoacoustic model work with Speex in wideband operation. It was found that the psychoacoustic model in the narrowband encoder is also suitable for the lower band of wideband speech. For the high-band, the reference psychoacoustic model is used. Although it would be possible to compute the masking curve on the wideband signal and then divide the spectrum in two bands, the added complexity (code and CPU time) outweighs the potential benefits.

5. EVALUATION AND RESULTS

We compare the reference Speex encoder to the modified encoder using the Vorbis psychoacoustic model. The experiment is conducted using Speex version 1.1.12, available from http://www.speex.org/. The Vorbis psychoacoustic model can be enabled by configuring with --enable-vorbis-psy or defining the VORBIS_PSYCHO macro at compile time.

The Perceptual Evaluation of Speech Quality (PESQ) tool, as defined by ITU-T recommendation G.862.2 [8, 9], is used to compare the encoders at various bitrates for both narrowband and wideband speech. While not a subjective mean opinion score (MOS) test, we consider the results to be meaningful because we are only comparing different noise-weighting filters for the same codec. The reference Speex decoder is used for both encoders. The test set is composed of 354 speech samples from 177 different speakers (87 male and 90 female) in 20 different languages taken from the NTT multi-lingual speech database. The Speex codec reference implementation supports variable search complexity. For the evaluation, the Speex variable complexity option is set to

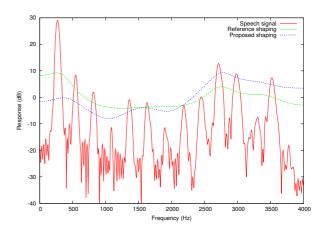


Figure 7: Noise shaping for reference encoder and modified encoder on a voiced frame. Arbitrary y-axis offset.

3, meaning that 3 simultaneous hypotheses are updated when searching for the best adaptive and fixed codebook entry.

Results for narrowband speech are presented in Fig. 8 with the AMR-NB, iLBC and GSM-FR codecs included as reference. Very low Speex bit-rates (2.15 and 4 kbps) are not included because it was found that the new model did not improve quality at those bit-rates. It can be observed that the proposed noise weighting significantly increases quality, especially at higher bit-rates. Also, it is worth noting that the improved encoder at 11 kbps achieves the same level of quality as the reference AMR-NB codec at 12.2 kbps.

In Fig. 9, the quality of the original and improved encoders are plotted with scaling of the x-axis by 5% and 20% for the original encoder. It can be observed from there that the improvement is equivalent to a bit-rate reduction of 5% at low bit-rates and up to 20% at high bit-rates.

Results for wideband speech are presented in Fig. 10 with the AMR-WB codec included as reference. Again, very low Speex bit-rates (3.95 and 5.8 kbps) are not included, since the quality was not improved. The proposed noise weighting significantly increases quality for wideband, clearly surpassing the quality of AMR-WB for bit-rates above 12.8 kbps.

In a last experiment, we evaluate the effect of complexity reduction methods proposed in Section 4.2. Only the

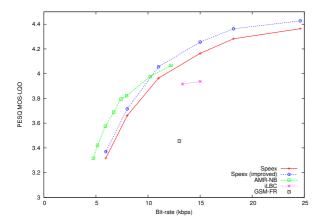


Figure 8: Quality of narrowband speech with and without improved noise model.

Complexity	PESQ MOS
proposed	4.256
C1	4.252
C2	4.259
C3	4.246
reference	4.164

Table 1: Impact of complexity reductions on speech quality. C1, C2 and C3 refer to the methods described in Section 4.2.

15 kbps narrowband mode is evaluated using each of the three complexity reduction methods proposed. Results in Table 1 show that methods C1 and C2 have no significant impact on quality, while C3 only has a small negative impact. This makes C3 an attractive choice, since it could lead to quality improvements without increasing the encoder complexity.

6. DISCUSSION AND CONCLUSION

In this work, we have demonstrated how to improve the quality of a CELP codec by choosing a better psychoacoustic model. It has been observed that the improvement is more significant at high bit-rate. We find this result unintuitive because we expect it to be easier to improve a lower quality modes. One hypothesis we propose to explain this result is the fact that at higher bit-rates, there are more bits available and thus more possibilities to change the bit allocation based on the noise weighting.

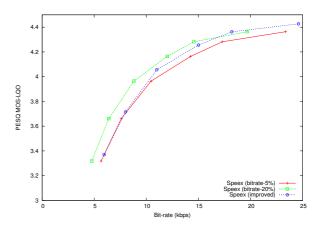


Figure 9: Bit-rate equivalent of the quality improvement.

Another hypothesis is that the lower the bit-rate, the further the optimal noise shaping is from the ideal masking curve we use.

Unlike the work by [10], this proposed improvement to Speex can be made without affecting compatibility or requiring any modification on the decoder side. Also, because it only applies to the encoder side, the technique could also improve the quality of other existing and future CELP codecs.

We believe this work clearly demonstrates that the noise weighting currently used in CELP codecs has become inadequate. We have also shown that improved noise weighting does not necessarily require increasing the complexity of the encoder. For these reasons, it would be desirable to further investigate alternative noise weighting filters for use in CELP. We would also like to validate the results in this paper with a subjective MOS test.

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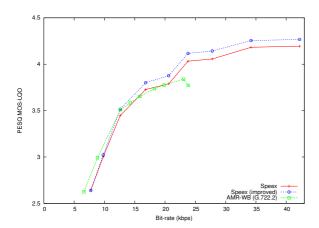


Figure 10: Quality of wideband speech with and without improved noise model.

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